

Guard-free OFDM transmission for Underwater Acoustic Communications



UNIVERSITY *of York*

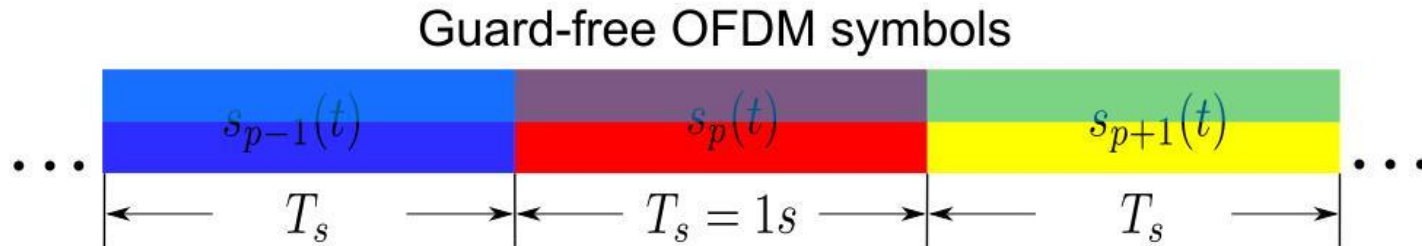
Yuriy Zakharov

yury.zakharov@york.ac.uk

Overview

- ❖ Data transmission
- ❖ Sea trials / channel model
- ❖ Receiver design:
 - ✓ Single-hydrophone receiver
 - ✓ Multiple-antenna receiver
 - ✓ Doppler estimation
- ❖ Conclusions

Transmitted signal

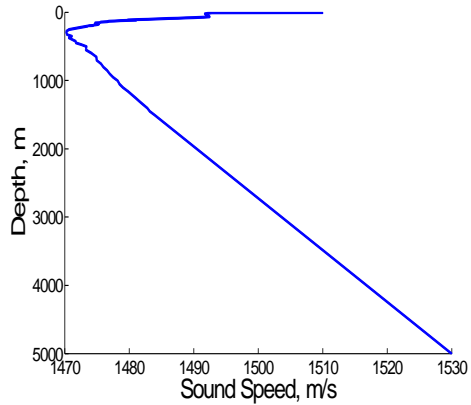
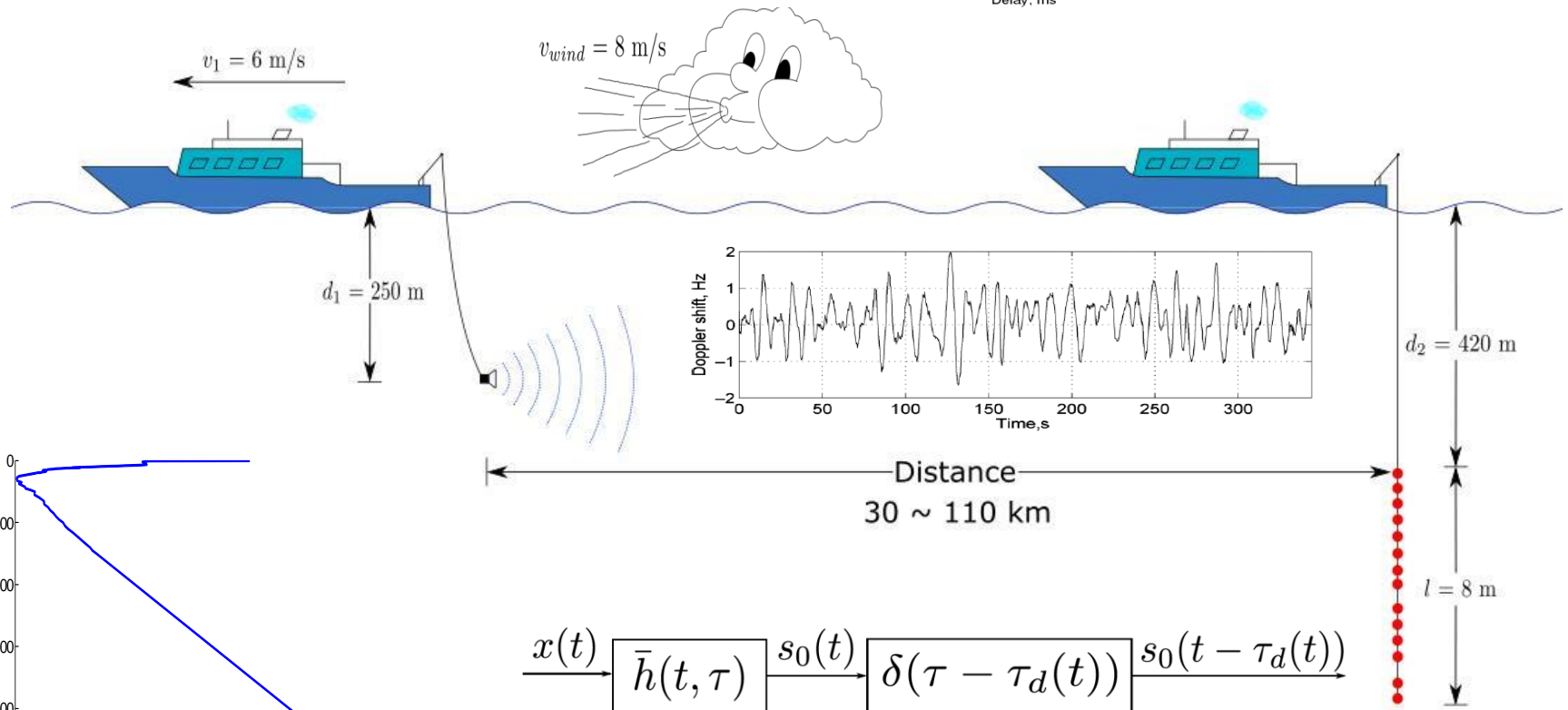
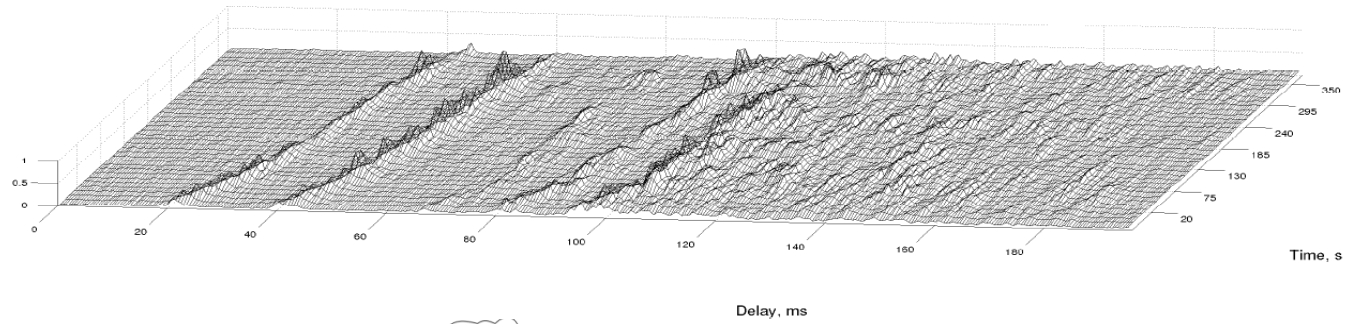
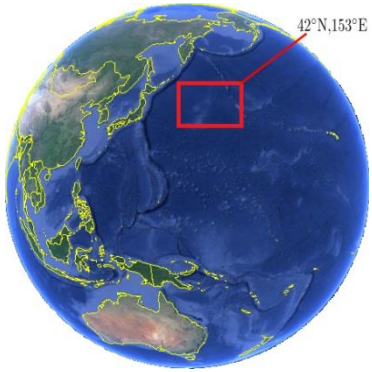


$$s_p(t) = \Re \left\{ e^{j2\pi f_c t} \sum_{k=-\frac{N_s}{2}}^{\frac{N_s}{2}} [M(k) + jD_p(k)] e^{\frac{j2\pi kt}{T_s}} \right\} = d(t) + p(t)$$

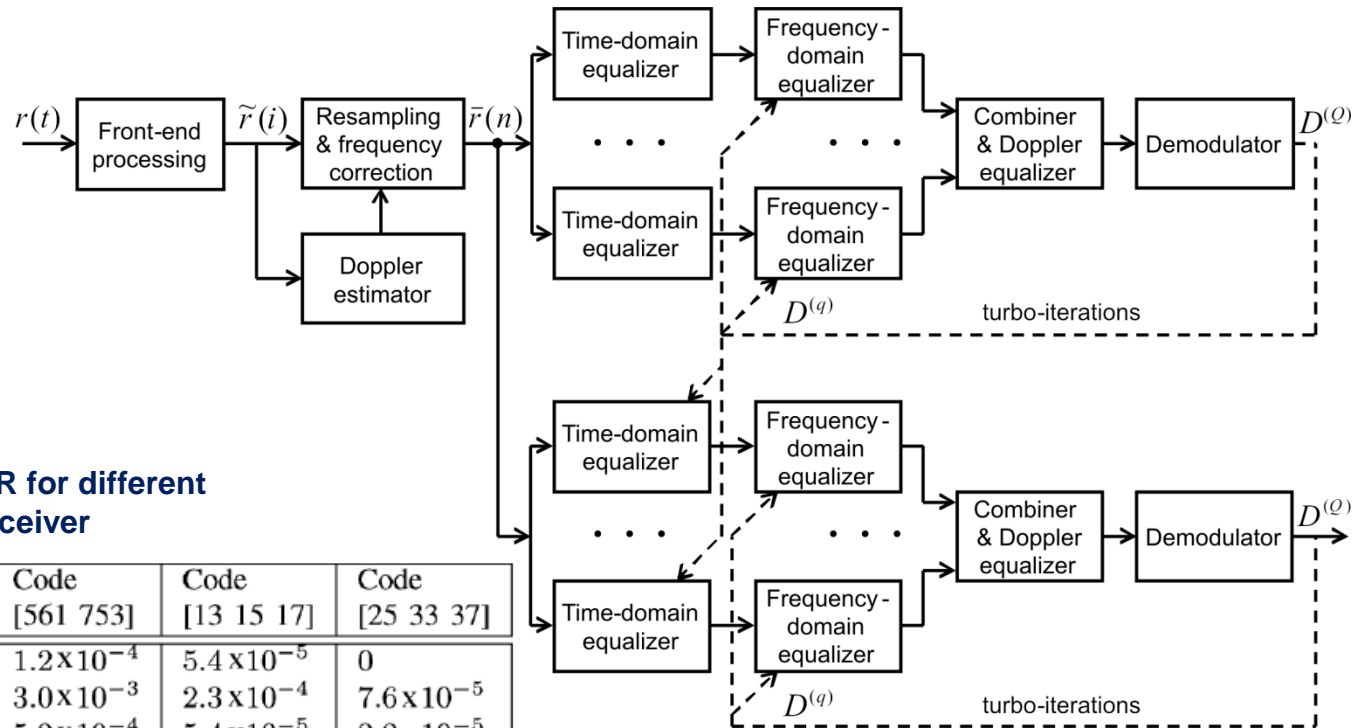
- $f_c = 3072$ Hz
- $N_s = 1024$: number of sub-carriers
- $F = 1024$ Hz: frequency bandwidth
- $M(k) \in [-1, +1]$: pilot sequence
- $D_p(k) \in [-1, +1]$: information data (coded across sub-carriers)
- $p = 1, \dots, L$: index of an OFDM symbol

Superimposed
data and pilot signals

Sea trials



Structure of single-hydrophone receiver

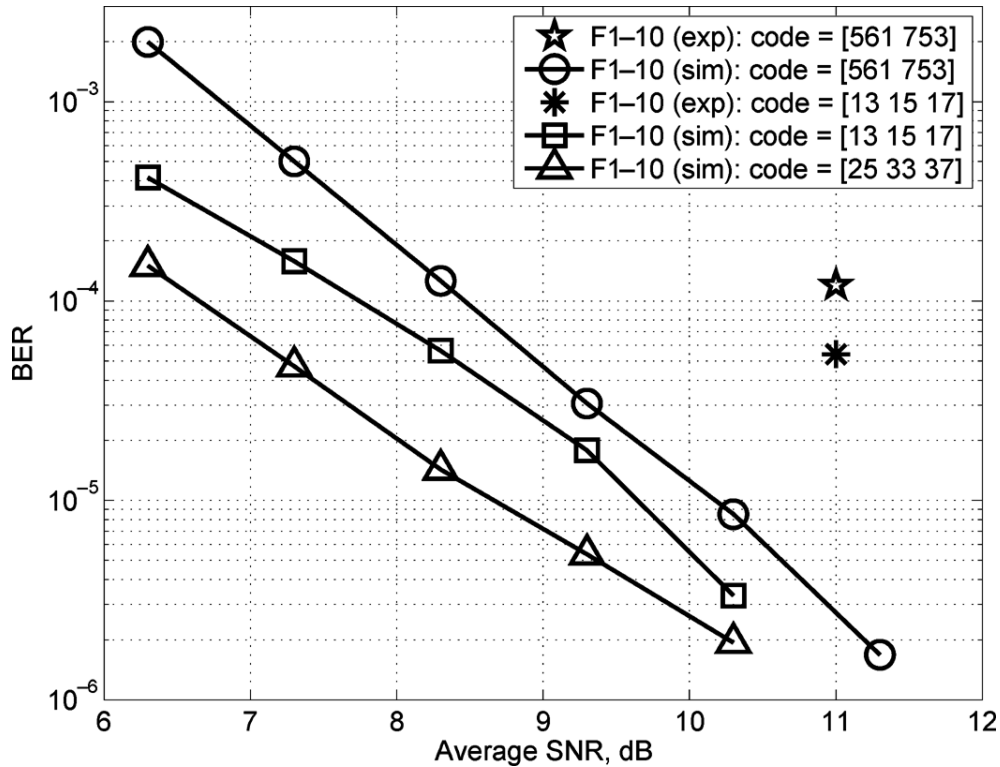


Session F1-10 (79 – 81 km): BER for different configurations of the receiver

Receiver configuration	Code [561 753]	Code [13 15 17]	Code [25 33 37]
1. Full receiver (av. SNR = 11 dB)	1.2×10^{-4}	5.4×10^{-5}	0
2. No oversampling $N_\tau = 1$	3.0×10^{-3}	2.3×10^{-4}	7.6×10^{-5}
3. Oversampling $N_\tau = 2$	5.2×10^{-4}	5.4×10^{-5}	2.2×10^{-5}
4. No fine Doppler estimation	1.9×10^{-4}	7.6×10^{-5}	3.2×10^{-5}
5. No turbo iterations ($Q = 1$)	3.1×10^{-4}	5.9×10^{-5}	3.2×10^{-5}
6. No Doppler equalizer ($L_d = 1$)	2.7×10^{-4}	5.4×10^{-5}	0
7. Zero-order interpolation	7.9×10^{-4}	1.2×10^{-4}	5.4×10^{-5}
8. Linear interpolation	2.1×10^{-4}	5.4×10^{-5}	0
9. Not accurate equalizer delay	1.0×10^{-3}	2.5×10^{-4}	7.6×10^{-5}
10. No frequency-domain equalizer	2.9×10^{-4}	6.5×10^{-5}	0
11. Sparse channel estimation	7.8×10^{-5}	5.4×10^{-5}	0

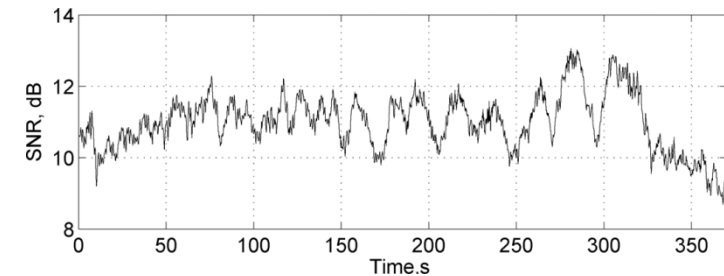
Y. Zakharov and A. Morozov, "OFDM transmission without guard interval in fast-varying underwater acoustic channels." *IEEE Journal of Oceanic Engineering* (2015).

Comparison with an *ideal* receiver

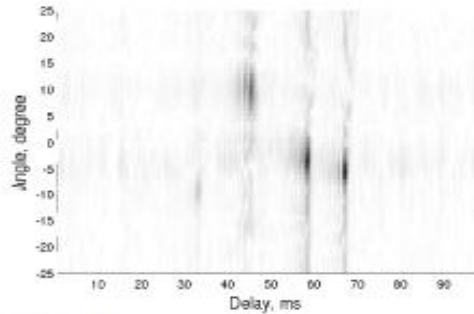


Ideal receiver:

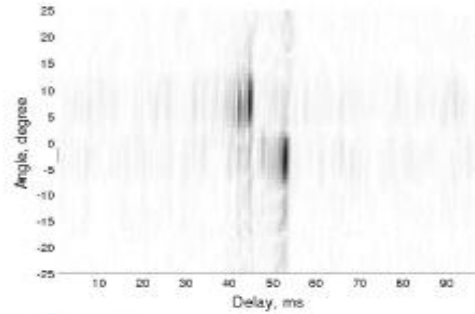
1. The channel is perfectly known.
2. The same multipath delay spread as in the sea experiment.
3. There is a cyclic prefix longer than the channel delay spread, no ISI.
4. No Doppler spread, no ICI.
5. No superimposed pilot, thus no interference from the pilot.
6. The SNR time-variation is the same as in the sea experiment



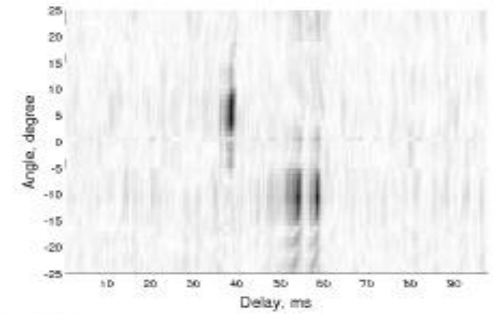
Space-time clustering



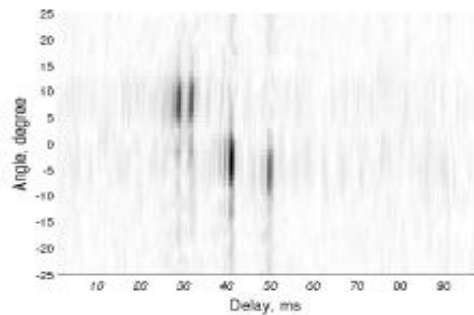
(d) 50 km



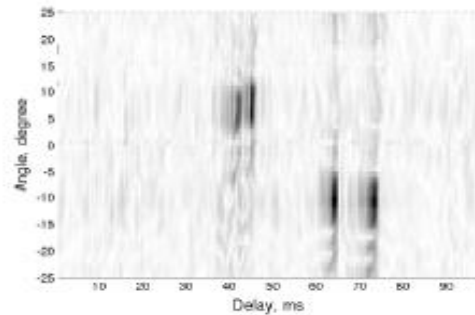
(e) 94 km



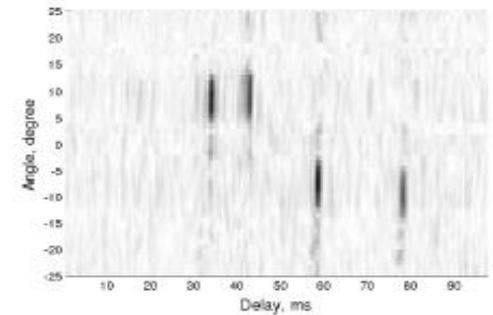
(f) 100 km



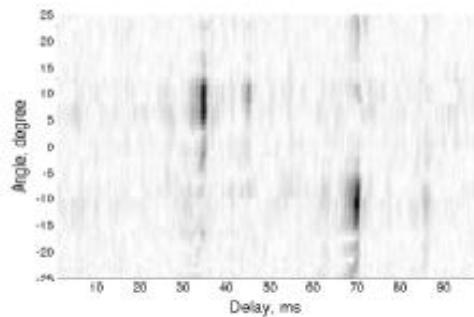
(g) 102 km



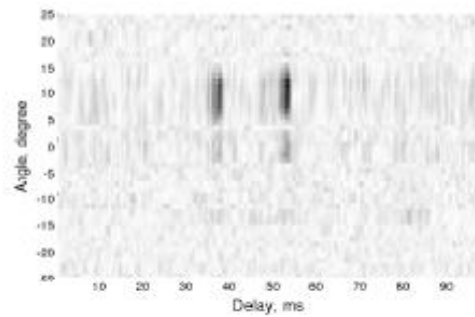
(h) 103 km



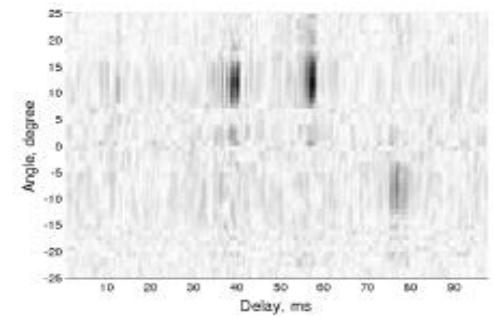
(i) 105 km



(j) 106 km



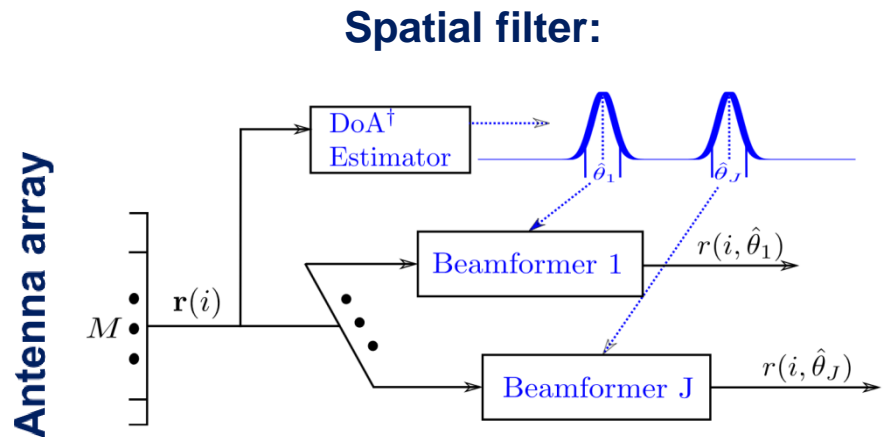
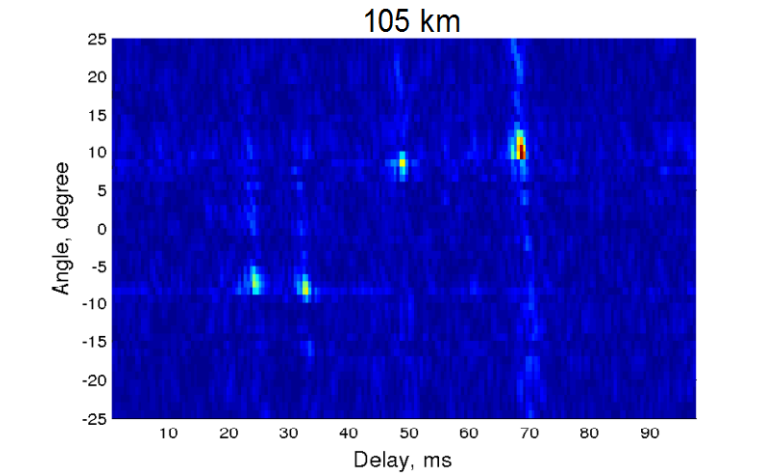
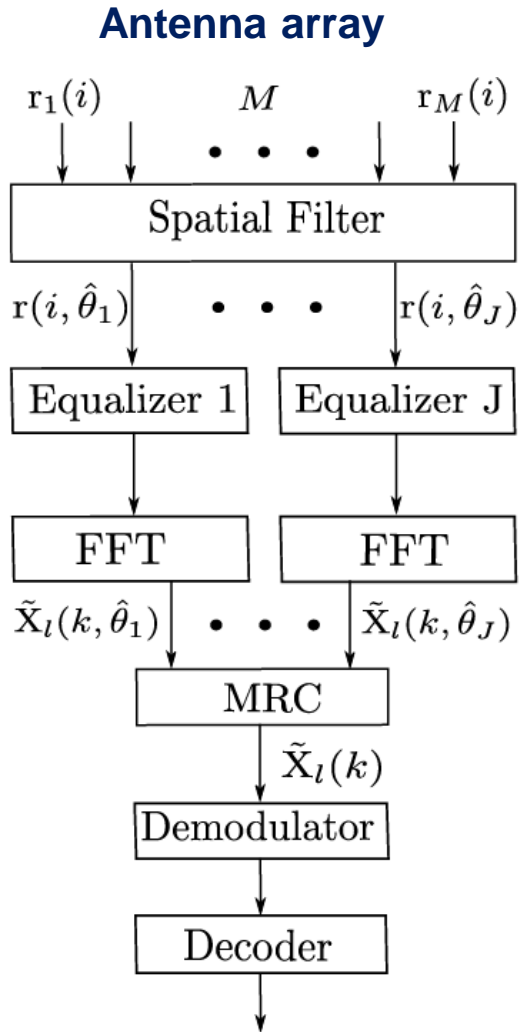
(k) 109 km



(l) 110 km

J. Li and Y. Zakharov, "Efficient use of space-time clustering for underwater acoustic communications", *IEEE Journal of Oceanic Engineering*, (2017 to appear).

Receiver with space-time processing

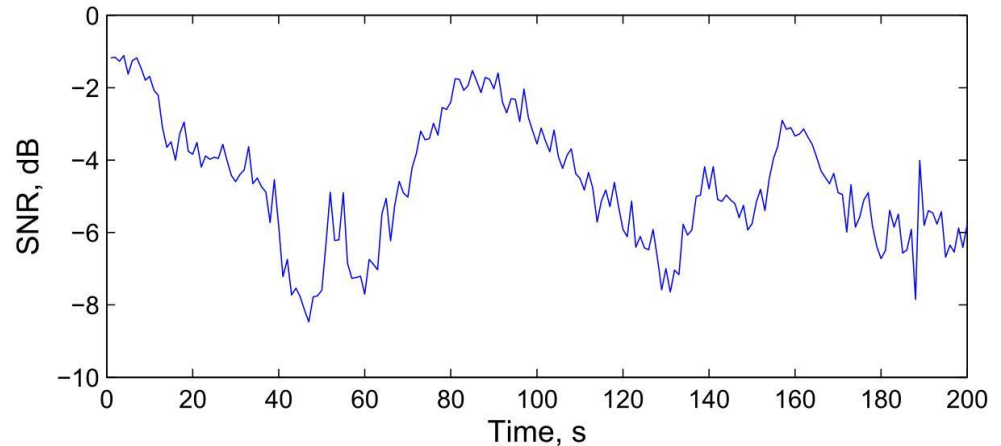


BER performance (105 km)

Receiver	BER(code 1/3)	Complexity(MACs) [†]
Single hydrophone	0.45	83×10^6
All 14 hydrophones	2.7×10^{-3}	1160×10^6
Single angle 8.4°	9.1×10^{-2}	100×10^6
Two angles with max powers	8.9×10^{-2}	185×10^6
Clusters (angles 8.4° and -9°)	0	185×10^6

code rate : 1/3 [225 331 367]

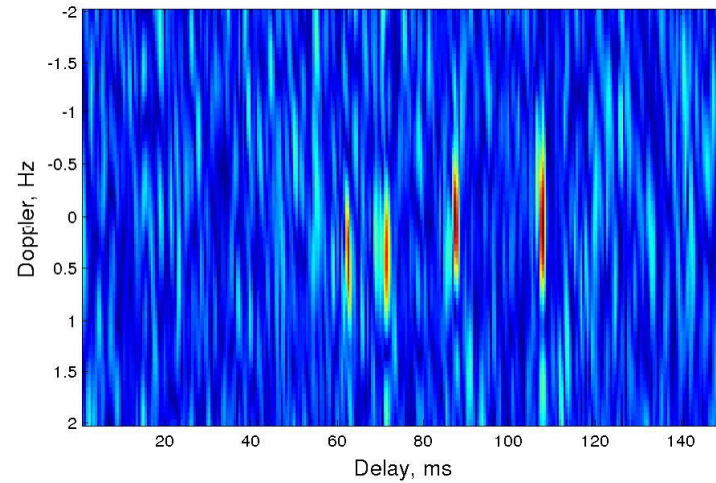
Single-hydrophone SNR:



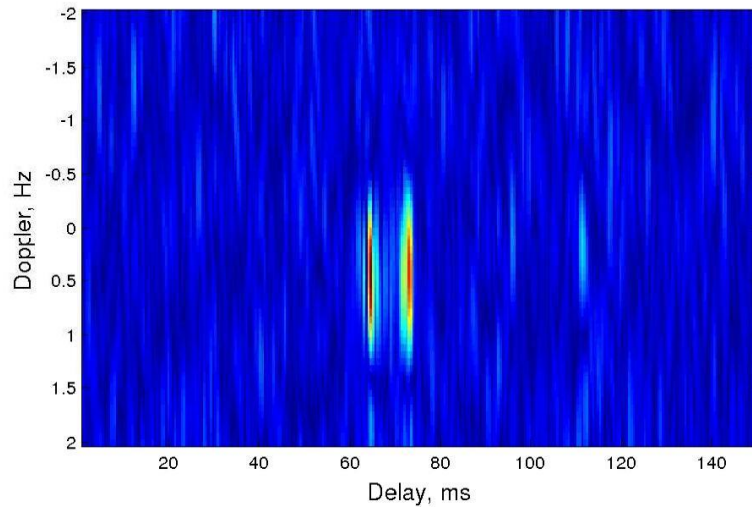
†: multiply-accumulate operations

Reducing the delay and Doppler spread (105 km)

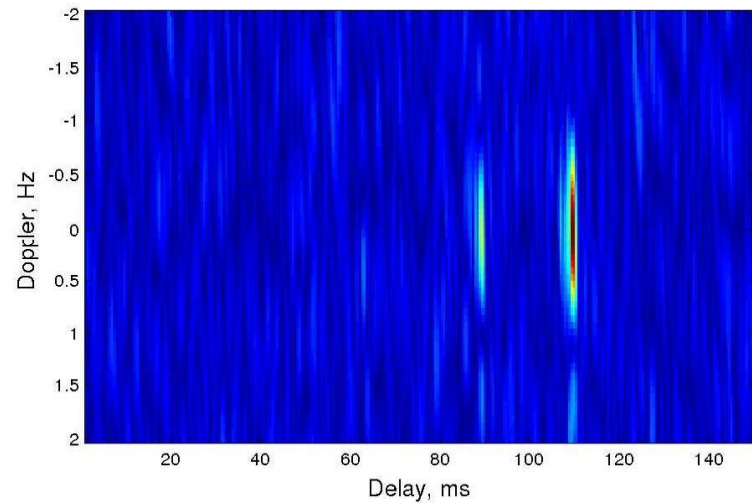
Hydrophone 1:



-9° :



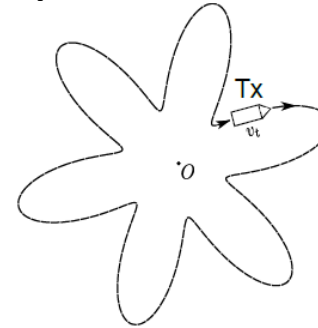
8.4° :



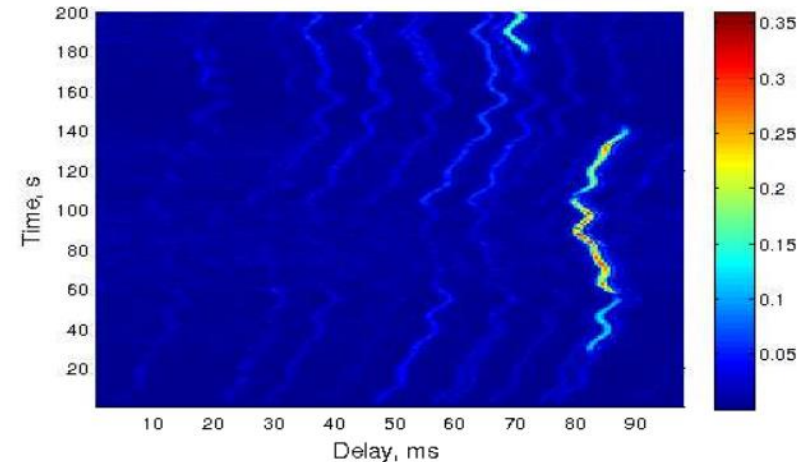
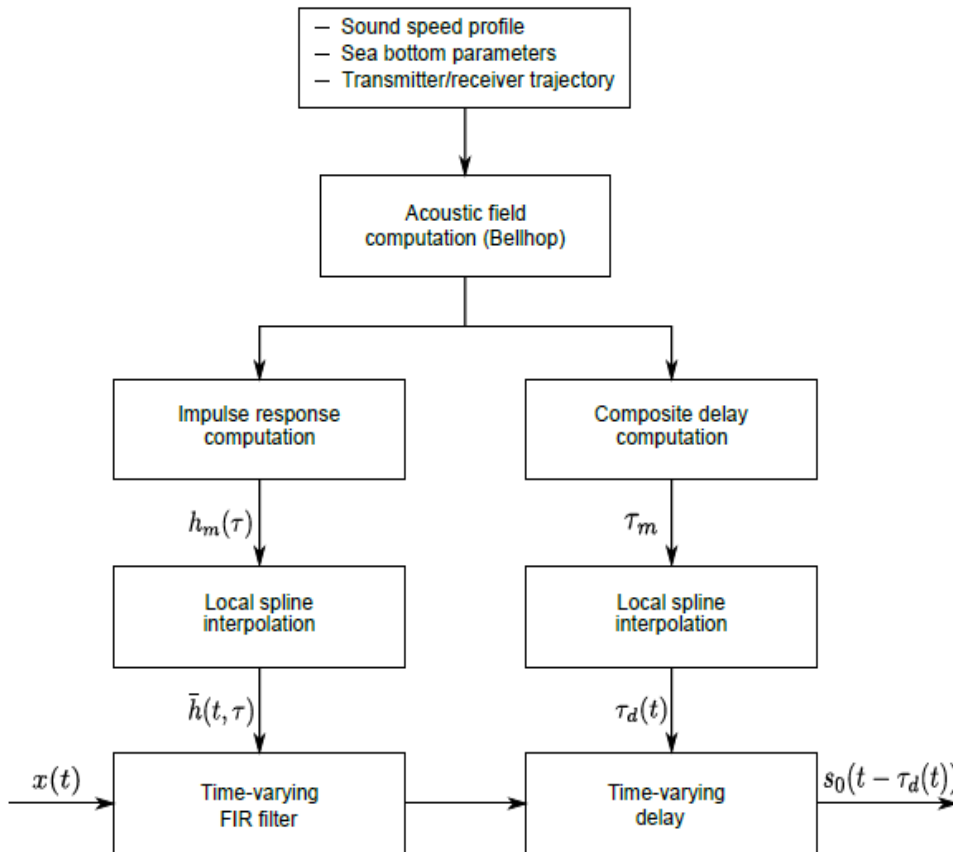
Waymark model: Virtual signal transmission

$$x(t) \rightarrow \bar{h}(t, \tau) \rightarrow s_0(t) \rightarrow \delta(\tau - \tau_d(t)) \rightarrow s_0(t - \tau_d(t))$$

For example:



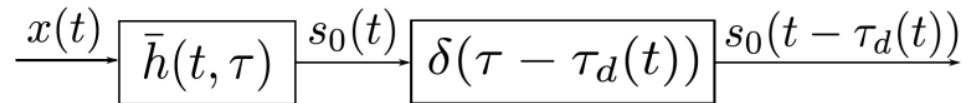
(c) Flower circle movement of the transmitter; O is the center of the flower.



(d) Summer SSP, flower circle movement (5 km).

C. Liu, Y. Zakharov, and T. Chen, "Doubly selective underwater acoustic channel model for a moving transmitter/receiver." *IEEE Transactions on Vehicular Technology* (2012).

Multi-branch autocorrelation (MBA) Doppler estimator



BER PERFORMANCE OF THE RECEIVER WITH THE THREE DOPPLER ESTIMATORS; DATA RATE: 1/2 BPS/HZ.

Doppler estimator	Code [3 7]	Code [23 35]	Code [561 753]
CAF	$4.5 \cdot 10^{-3}$	$8.5 \cdot 10^{-4}$	$2.0 \cdot 10^{-5}$
SBA	0.30	0.34	0.37
MBA	$4.8 \cdot 10^{-3}$	$9.2 \cdot 10^{-4}$	0

CAF: cross-ambiguity function

SBA: single-branch autocorrelation

MBA: multi-branch autocorrelation

Comparing to the CAF method, the reduction in complexity is about 10 times.

J. Li, Y. Zakharov, and B. Henson, "Multi-branch autocorrelation method for Doppler estimation in UWA channels", *IEEE Journal of Oceanic Engineering*, (2017, under review).

Summary

- ❖ **Guard-free OFDM transmission makes sense in UWA channels**
- ❖ **Superimposed pilot signals are very useful in dynamic UWA channels.**
- ❖ **Space-time processing significantly improves the detection performance; the cluster combining takes into account the specific UWA propagation and also very useful.**
- ❖ ***Waymark* model allows the virtual signal transmission. It is useful to make complicated “sea experiments”.**
- ❖ **MBA Doppler estimator has performance close to that of the ambiguity function Doppler estimator, but with significantly reduced complexity.**